Fixed-time Synchronization of Networked Uncertain Euler-Lagrange Systems

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Abstract—This paper considers the fixed-time control problem of a multi-agent system composed of a class of Euler-Lagrange dynamics with parametric uncertainty and a dynamic leader under a directed communication network. A distributed fixed-time observer is first proposed to estimate the desired trajectory and then a fixed-time controller is constructed by transforming uncertain Euler-Lagrange systems into second-order systems and utilizing the backstepping design procedure. The overall design guarantees that the synchronization errors converge to zero in a prescribed time independent of initial conditions. The control design conditions can also be relaxed for a weaker finite-time control requirement.

Index Terms—Finite-time control, fixed-time control, multi-agent systems, Euler-Lagrange systems, directed graph

I. INTRODUCTION

Fixed-time control for multi-agent systems, requiring exact achievement of a collective behavior in a prescribed time independent of initial conditions, or finite-time control of a weaker requirement allowing the prescribed time dependent on initial conditions, has attracted researchers' extensive attention over the past years due to its potential advantages in transient performance and robustness property [1]. The early work on finite-time formation control of single-integrator multi-agent systems can be found in [2]. For the leader-following consensus problem of general linear multi-agent systems, [3] proposed two classes of finite-time observers to estimate the second-order leader dynamics, which can work in undirected and directed communication networks, respectively. More efforts have also been devoted to nonlinear systems. For example, [4] considered the finite-time control of first-order multi-agent systems with unknown nonlinear dynamics, while both first-order and second-order nonlinear systems were considered in [5]. In particular, observerbased control was proposed to solve the leader-following fixedtime consensus problem under the strongly connected communication network. The fixed-time consensus problem was also investigated for double-integrator systems under directed communication network and more general multi-agent systems with high-order integrator dynamics in [6], [7], respectively.

Euler-Lagrange systems capture a large class of contemporary engineering problems and finite-time control of this class of systems has been intensively investigated, especially in the individual setting. For example, [8] considered finite-time control for an Euler-Lagrange system based on the method for a double-integrator system, while [9], [10] further dealt with nonlinear systems in the presence of uncertainties. The work in [11] studied a non-singular sliding surface and constructed a continuous finite-time control strategy for uncertain

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Z. Chen is with with the School of Electrical Engineering and Computing, The University of Newcastle, Callaghan, NSW 2308, Australia. E-mail: zhiyong.chen@newcastle.edu.au Euler-Lagrange system. Furthermore, [12] designed an adaptive controller to track a desired trajectory in finite time and [13] proposed a method for handing both uncertain dynamics and globally unbounded disturbances.

The research on fixed-time or finite-time control of uncertain Euler-Lagrange systems in a network setting is relatively rare. Some related results can be found in [14] where, by adaptive control technique, a finite-time synchronization controller was constructed for a multiagent system modeled by some mechanical nonlinear systems with a connected communication network. The recent work reported in [15] studied finite-time coordination behavior of a multiple Euler-Lagrange system with an undirected network in the absence of uncertainties. In particular, with the introduction of auxiliary variables, the system can be converted into a simpler form such that the adding a power integrator method can be applied to ensure the convergence.

This paper provides a solution to the leader-following fixed-time synchronization problem for multiple Euler-Lagrange systems with parametric uncertainty. The strategy is based on a class of observers that can accurately estimate a dynamic trajectory in a fixed time. The design relaxes the undirected and connected assumption for the communication network in [5], [7], [14], [15] and considers a directed network graph. Then an observer-based controller is proposed for the multi-agent system composed of a dynamic leader and multiple heterogeneous Euler-Lagrange dynamics, as opposed to the finitetime control method for multiple special mechanical systems in [14]. In particular, the distributed control law is able to guarantee each Euler-Lagrange system can track a desired trajectory in a prescribed time, independent of initial conditions. It is worth mentioning that the control design conditions can be relaxed for a weaker finite-time control requirement. Also, a reduced continuous controller can be directly applied to the fixed-time synchronization problem for secondorder nonlinear systems with a directed graph.

Throughout the paper, we use the following notations. For a vector $x = [x_1, \dots, x_n]^T \in \mathbb{R}^n$, $||x||_1 = |x_1| + \dots + |x_n|$ represents its Manhattan (\mathcal{L}_1) norm, $||x|| = \sqrt{x_1^2 + \dots + x_n^2}$ its Euclidean (\mathcal{L}_2) norm, and $|x| = [|x_1|, \dots, |x_n|]^T$ its element-wise absolute valued vector. For a matrix X, |X| is also defined as its element-wise absolute valued matrix. The power function operator is element-wise in terms of $x^a = [x_1^a, \dots, x_n^a]^T$ for a > 0. For two vectors (matrices) X and Y, comparison operators are element-wise; for example, $X \ge Y$ means $x_{ij} \ge y_{ij}$ for every x_{ij} and y_{ij} , the (i, j)-elements of X and Y, respectively. The operator sig^a $(x) = [\text{sign}(x_1)|x_1|^a, \dots, \text{sign}(x_n)|x_n|^a]^T$ is defined for a > 0 and the sign function sign(\cdot).

II. PROBLEM FORMULATION

Consider a group of m-link robotic manipulators of the following Euler-Lagrange dynamics

$$M_i(q_i)\ddot{q}_i + C_i(q_i, \dot{q}_i)\dot{q}_i + G_i(q_i) = \tau_i, \ i = 1, \cdots, N,$$
(1)

where $q_i \in \mathbb{R}^m$, $\dot{q}_i \in \mathbb{R}^m$ are the vectors of generalized position and velocity of the *i*-th robotic manipulator, also called agent *i*, $M_i(q_i) \in \mathbb{R}^{m \times m}$ is a symmetric and positive definite inertia matrix, $C_i(q_i, \dot{q}_i)\dot{q}_i \in \mathbb{R}^m$ contains the Coriolis and centrifugal forces, $G_i(q_i) \in \mathbb{R}^m$ is the gravitational torque, and $\tau_i \in \mathbb{R}^m$ is the vector of control force. The reference is generated by a leader system, called agent 0, described as follows:

$$\dot{\eta}_0 = S\eta_0, \ q_0 = E\eta_0,$$
 (2)

where $\eta_0 \in \mathbb{R}^n$ is the state, $q_0 \in \mathbb{R}^m$ is the desired trajectory to track, and $S \in \mathbb{R}^{n \times n}$, $E \in \mathbb{R}^{m \times n}$ are constant matrices.

The multi-agent system under consideration is composed of the N dynamics in (1) and the dynamic leader (2). The information flow among all the N + 1 agents is described by a digraph $\mathcal{G} = (\mathcal{V}, \mathcal{E})$ where $\mathcal{V} = \{0, 1, \dots, N\}$ is the node set and $\mathcal{E} \subset \mathcal{V} \times \mathcal{V}$ is the edge set. Each element $(j, i) \in \mathcal{E}$ represents the edge from agent j to agent i. For $i, j \in \mathcal{V}$, $a_{ii} = 0$, $a_{ij} > 0$ if $(j, i) \in \mathcal{E}$ and $a_{ij} = 0$ otherwise. Let $H = [h_{ij}]_{i,j=1}^N$ be the Laplacian matrix of the subnetwork composed of agents $1, \dots, N$, where $h_{ii} = \sum_{j=0}^N a_{ij}$ and $h_{ij} = -a_{ij}$ for $i \neq j, i, j = 1, \dots, N$.

The objective of this paper is to design a distributed control law such that each agent of (1) can track the desired trajectory q_0 in fixed time. More specifically, we consider the class of control laws of the form

$$\tau_{i} = f_{1i}(q_{i}, \dot{q}_{i}, \eta_{i}, \dot{\eta}_{i})$$

$$\dot{\eta}_{i} = f_{2i}(\eta_{i}, \sum_{j=0}^{N} a_{ij}(\eta_{i} - \eta_{j})), \ i = 1, \cdots, N.$$
(3)

Let

$$x_{i} = \begin{bmatrix} q_{i} - q_{0} \\ \dot{q}_{i} - \dot{q}_{0} \\ \eta_{i} - \eta_{0} \end{bmatrix}, i = 1, \cdots, N, \ x = \begin{bmatrix} x_{1} \\ \vdots \\ x_{N} \end{bmatrix} \in \mathbb{R}^{n_{x}}.$$

Suppose the closed-loop system composed of (1), (2) and (3) possesses unique solutions in forward time for all initial conditions. Then the fixed-time synchronization problem can be described as follows based on the concept of fixed-time stability [16].

Fixed-time synchronization problem: Given the system composed of (1) and (2) with the corresponding digraph \mathcal{G} , design a distributed control law of the form (3) such that, for all initial conditions $x(0) = x_0$ and $\eta_0(0) = \eta_{00}$, the equilibrium x = 0of the closed-loop system is (globally) fixed-time stable. That is, the solution x(t) exists for $t \ge 0$ and x = 0 is Lyapunov stable, and moreover, there exists a fixed time T^* , independent of x_0 or η_{00} , such that

$$\lim_{t \to T^*} x(t) = 0,$$

$$x(t) = 0, \ t \ge T^*, \ \forall x_0 \in \mathbb{R}^{n_x}, \eta_{00} \in \mathbb{R}^n.$$
(4)

Remark 2.1: When the existence of a fixed time T^* is relaxed to the existence of a settling-time function $T(x_0, \eta_{00})$, the above fixedtime synchronization problem is called a finite-time synchronization problem, which is based on the definition of finite-time stability [1]. In practice, there is similarity between finite-time convergence and asymptotical (exponential) convergence, both of which require the convergence of trajectories to (proximity of) an equilibrium point in a finite amount of time which depends on the initial conditions. But fixed-time convergence is a more practically interesting feature which requests it happen in a prescribed time independent of initial conditions. There are more constraints on the controller design conditions that will be studied in this paper.

For the solvability of the aforementioned problem, we need the following standard assumption on the communication network.

Assumption 2.1: The graph \mathcal{G} contains a spanning tree with node 0 as the root.

Remark 2.2: Under Assumption 2.1, all the eigenvalues of H have positive real parts; see, e.g., [17]. By Theorem 2.5.3 of [18], there exists a positive definite diagonal matrix $\overline{D} \in \mathbb{R}^{N \times N}$ such that $H^T \overline{D} + \overline{D}H$ is positive definite. Let $\lambda_m > 0$ be the smallest eigenvalue of $H^T \overline{D} + \overline{D}H$ and $D = \text{diag}(d_1, \cdots, d_N) = 2\overline{D}/\lambda_m$. One has $H^T D + DH \ge 2I_N$.

We end this section with some technical lemmas from, e.g., [6], [16], [19] and [20], which will be used in the proofs of the main results in this paper.

Lemma 2.1: For any $\xi_i \in \mathbb{R}$, $i = 1, \dots, n$, and any $p \in (0, 1]$, $(\sum_{i=1}^{n} |\xi_i|)^p \leq \sum_{i=1}^{n} |\xi_i|^p \leq n^{1-p} (\sum_{i=1}^{n} |\xi_i|)^p$ [19]. For any p > 1, $\sum_{i=1}^{n} |\xi_i|^p \leq (\sum_{i=1}^{n} |\xi_i|)^p \leq n^{p-1} \sum_{i=1}^{n} |\xi_i|^p$ [6]. **Lemma 2.2:** [19] The inequality $|\xi_i^p - \xi_j^p| \leq 2^{1-p} |\xi_i - \xi_j|^p$ holds

Lemma 2.2: [19] The inequality $|\xi_i^p - \xi_j^p| \le 2^{1-p} |\xi_i - \xi_j|^p$ holds for $\forall \xi_i, \xi_j \in \mathbb{R}$ and 0 and <math>p is a ratio of two odd integers. **Lemma 2.3:** [20] The inequality $|\xi_i|^c |\xi_j|^d \le \frac{c}{c+d} r |\xi_i|^{c+d} + \frac{d}{c+d} r^{-\frac{c}{d}} |\xi_j|^{c+d}$ holds for $\forall \xi_i, \xi_j \in \mathbb{R}$ and c, d, r > 0.

Lemma 2.4: (Lemma 1, [16]) Consider the system $\dot{z} = \phi(z,t)$ where $\phi : \mathbb{R}^l \times [0,\infty) \mapsto \mathbb{R}^l$ satisfies $\phi(0,t) = 0$. Suppose there exits a continuously differentiable function $V : \mathbb{R}^l \mapsto \mathbb{R}$ such that (i) V is positive definite and proper; and (ii) there exist real numbers $p_0, q_0, p, q, k > 0$ with pk < 1 and qk > 1 such that $\dot{V}(z) \leq -(p_0(V(z))^p + q_0(V(z))^q)^k$. Then, the equilibrium z = 0is (globally) fixed-time stable and there is a constant settling-time $T^* \leq \frac{1}{p_0^k(1-pk)} + \frac{1}{q_0^k(qk-1)}$.

III. DISTRIBUTED OBSERVER DESIGN

As the agents not connected to agent 0 do not have access to the information of the dynamic leader (2), its state needs to be estimated by a properly designed fixed-time observer as follows:

$$\dot{\eta}_{i} = S\eta_{i} - c_{1}y_{i} - c_{2}\operatorname{sig}^{a}(y_{i}) - c_{3}\operatorname{sig}^{b}(y_{i}),$$

$$y_{i} = \sum_{j=0}^{N} a_{ij}(\eta_{i} - \eta_{j}), \ i = 1, \cdots, N.$$
(5)

In this section, we construct a lemma based on the fixed-time observer (5). Let $\eta^T = [\eta_0^T, \eta_1^T, \cdots, \eta_N^T]^T$ for the convenience of presentation.

Lemma 3.1: Consider the system composed of (2) and (5) under Assumption 2.1 with 0 < a < 1, $b > \frac{1}{a} > 1$, $c_1 > ||D \otimes S||$ and $c_2, c_3 > 0$. There exists a constant settling-time $T_1^* \ge 0$ such that, $\forall \eta(0) \in \mathbb{R}^{(N+1)n}$,

$$\lim_{t \to T_1^*} (\eta_i(t) - \eta_0(t)) = 0,$$

$$\eta_i(t) - \eta_0(t) = 0, \ t \ge T_1^*, \ i = 1, \cdots, N.$$
(6)

Proof: Let $\bar{\eta}_i = \eta_i - \eta_0$, $i = 0, 1, \dots, N$. The observer (5) can be rewritten as

$$\dot{\bar{\eta}}_i = S\bar{\eta}_i - c_1 y_i - c_2 \operatorname{sig}^a(y_i) - c_3 \operatorname{sig}^b(y_i),$$

$$y_i = \sum_{j=0}^N a_{ij}(\bar{\eta}_i - \bar{\eta}_j), \ i = 1, \cdots, N.$$
 (7)

Let $Y_i = c_1 y_i + c_2 \operatorname{sig}^a(y_i) + c_3 \operatorname{sig}^b(y_i)$ and $\overline{\eta}, y, Y$ be the column stacks of $\overline{\eta}_i, y_i, Y_i, i = 1, \dots, N$. Note $y = (H \otimes I_n)\overline{\eta}$ and

$$\dot{y} = (I_N \otimes S)y - (H \otimes I_n)Y.$$
(8)

And, let

$$V(y) = \sum_{i=1}^{N} \left(\frac{c_2 d_i}{1+a} \| y_i^{1+a} \|_1 + \frac{c_3 d_i}{1+b} \| y_i^{1+b} \|_1 \right) \\ + \frac{c_1}{2} y^T (D \otimes I_n) y.$$
(9)

Along the trajectory of (8), the time derivative of V(y) satisfies

$$\begin{split} \dot{V}(y) &= \sum_{i=1}^{N} d_i (c_2 \operatorname{sig}^a(y_i) + c_3 \operatorname{sig}^b(y_i))^T \dot{y}_i + c_1 y^T (D \otimes I_n) \dot{y} \\ &= Y^T (D \otimes I_n) \dot{y} \\ &= Y^T (D \otimes S) y - \frac{1}{2} Y^T ((H^T D + DH) \otimes I_n) Y \\ &\leq \frac{1}{2} Y^T Y + \frac{1}{2} \|D \otimes S\|^2 y^T y - Y^T Y \\ &= -\frac{1}{2} (\|Y\|^2 - \|D \otimes S\|^2 \|y\|^2). \end{split}$$

Further calculation shows that

$$\begin{split} \|Y_i\|^2 &= \|c_1y_i + c_2\mathrm{sig}^a(y_i) + c_3\mathrm{sig}^b(y_i))\|^2 \\ &\geq c_1^2 \|y_i\|^2 + c_2^2 \|\mathrm{sig}^a(y_i)\|^2 + c_3^2 \|\mathrm{sig}^b(y_i)\|^2 \\ &\geq c_1^2 \|y_i\|^2 + c_2^2 \|y_i^{2a}\|_1 + c_3^2 \|y_i^{2b}\|_1. \end{split}$$

By Lemma 2.1, for 0 < a < 1 and b > 1,

$$\sum_{i=1}^{N} \|y_i^{2a}\|_1 \ge (\sum_{i=1}^{N} \|y_i^{2}\|_1)^a = (\|y\|^2)^a,$$
$$\sum_{i=1}^{N} \|y_i^{2b}\|_1 \ge \frac{1}{(nN)^{b-1}} (\|y\|^2)^b.$$

As a result,

$$||Y||^{2} = \sum_{i=1}^{N} ||Y_{i}||^{2} \ge c_{1}^{2} ||y||^{2} + c_{2}^{2} (||y||^{2})^{a} + \frac{c_{3}^{2}}{(nN)^{b-1}} (||y||^{2})^{b}$$

and hence

$$\dot{V}(y) \leq -\frac{1}{2}(c_1^2 - \|D \otimes S\|^2)\|y\|^2 - \frac{c_2^2}{2}(\|y\|^2)^a -\frac{c_3^2}{2(nN)^{b-1}}(\|y\|^2)^b \leq -\hat{c}_1\left(\|y\|^2 + (\|y\|^2)^a + (\|y\|^2)^b\right)$$
(10)

for

$$\hat{c}_1 = \min\left\{\frac{1}{2}(c_1^2 - \|D \otimes S\|^2), \frac{c_2^2}{2}, \frac{c_3^2}{2(nN)^{b-1}}\right\} > 0.$$

Analysis on (9) using Lemma 2.1 and noting $0 < \frac{2a}{1+a} < 1$ and $\frac{a(1+b)}{1+a} > 1$ gives

$$V^{\frac{2a}{1+a}} \leq \sum_{i=1}^{N} \left(\left(\frac{c_1 d_i}{2}\right)^{\frac{2a}{1+a}} \|y_i^{\frac{4a}{1+a}}\|_1 + \left(\frac{c_2 d_i}{1+a}\right)^{\frac{2a}{1+a}} \|y_i^{2a}\|_1 + \left(\frac{c_3 d_i}{1+b}\right)^{\frac{2a}{1+a}} \|y_i^{\frac{2a(1+b)}{1+a}}\|_1 \right)$$
$$\leq \hat{c}_2 \left(\left(\|y\|^2\right)^{\frac{2a}{1+a}} + \left(\|y\|^2\right)^a + \left(\|y\|^2\right)^{\frac{a(1+b)}{1+a}} \right)$$

for $d_{\max} = \max\{d_1, \cdots, d_N\}$ and

$$\hat{c}_{2} = \max\left\{ \left(\frac{c_{1}d_{\max}}{2}\right)^{\frac{2a}{1+a}} (nN)^{1-\frac{2a}{1+a}}, \\ \left(\frac{c_{2}d_{\max}}{1+a}\right)^{\frac{2a}{1+a}} (nN)^{1-a}, \left(\frac{c_{3}d_{\max}}{1+b}\right)^{\frac{2a}{1+a}} \right\}.$$

Since $a < \frac{2a}{1+a} < 1$ and $a < \frac{a(1+b)}{1+a} < b$, we can easily verify

$$(\|y\|^2)^{\frac{2a}{1+a}} \le \|y\|^2 + (\|y\|^2)^a (\|y\|^2)^{\frac{a(1+b)}{1+a}} \le (\|y\|^2)^a + (\|y\|^2)^b$$

and hence

$$V^{\frac{2a}{1+a}} \le \hat{c}_2 \left(\|y\|^2 + 3(\|y\|^2)^a + (\|y\|^2)^b \right).$$
(11)

Similarly, for b > 1, $\frac{2b}{1+b} > 1$ and $\frac{b(1+a)}{1+b} > 1$,

$$V^{\frac{2b}{1+b}} \le \hat{c}_3 \left((\|y\|^2)^{\frac{2b}{1+b}} + (\|y\|^2)^{\frac{b(1+a)}{1+b}} + (\|y\|^2)^b \right)$$

for

$$\hat{c}_3 = \max\left\{ \left(\frac{c_1 d_{\max}}{2}\right)^{\frac{2b}{1+b}}, \left(\frac{c_2 d_{\max}}{1+a}\right)^{\frac{2b}{b+1}}, \left(\frac{c_3 d_{\max}}{1+b}\right)^{\frac{2b}{b+1}} \right\} \times (3nN)^{\frac{b-1}{b+1}}.$$

Since $1 < \frac{2b}{1+b} < b$ and $a < \frac{b(1+a)}{1+b} < b$, we can easily verify

$$(\|y\|^2)^{\frac{2b}{1+b}} \le \|y\|^2 + (\|y\|^2)^b (\|y\|^2)^{\frac{b(1+a)}{1+b}} \le (\|y\|^2)^a + (\|y\|^2)^b$$

and hence

$$V^{\frac{2b}{1+b}} \le \hat{c}_3 \left(\|y\|^2 + (\|y\|^2)^a + 3(\|y\|^2)^b \right)$$
(12)

Finally, by (11) and (12), one has

$$\frac{1}{\hat{c}_2}V^{\frac{2a}{1+a}} + \frac{1}{\hat{c}_3}V^{\frac{2b}{1+b}} \le 4\left(\left\|y\right\|^2 + \left(\left\|y\right\|^2\right)^a + \left(\left\|y\right\|^2\right)^b\right),$$

which, compared with (10), implies

$$\dot{V} \leq -\frac{\hat{c}_1}{4\hat{c}_2}V^{\frac{2a}{1+a}} - \frac{\hat{c}_1}{4\hat{c}_3}V^{\frac{2b}{1+b}}.$$

By Lemma 2.4, the system (8) is fixed-time stable. In particular, there exists a constant

$$T_1^* \le \frac{4\hat{c}_2(a+1)}{\hat{c}_1(1-a)} + \frac{4\hat{c}_3(b+1)}{\hat{c}_1(b-1)},$$

such that $\lim_{t\to T_1^*} y(t) = 0$ and y(t) = 0, $t \ge T_1^*$. Under Assumption 2.1, we have $\bar{\eta} = (H^{-1} \otimes I_n)y$ and hence (6). The proof is thus completed.

Remark 3.1: When $c_3 = 0$, the observer (5) reduces to a finitetime observer

$$\dot{\eta}_{i} = S\eta_{i} - c_{1}y_{i} - c_{2}\operatorname{sig}^{a}(y_{i}),$$

$$y_{i} = \sum_{j=0}^{N} a_{ij}(\eta_{i} - \eta_{j}), \ i = 1, \cdots, N.$$
 (13)

Consider the system composed of (2) and (13) under Assumption 2.1 with 0 < a < 1, $c_1 > ||D \otimes S||$ and $c_2 > 0$. There exists a settlingtime function $T_1(\eta(0)) \ge 0$ such that, $\forall \eta(0) \in \mathbb{R}^{(N+1)n}$,

$$\lim_{t \to T_1} (\eta_i(t) - \eta_0(t)) = 0, \ i = 1, \cdots, N$$

$$\eta_i(t) - \eta_0(t) = 0, \ t \ge T_1(\eta(0)).$$
(14)

The proof of the above statement follows that of Lemma 3.1 using simple arguments. In particular, we can obtain

$$\dot{V} \le -\frac{\hat{c}_1}{3\hat{c}_2}V^{\frac{2a}{1+a}}$$

In other words, $\dot{V} + \rho V^{\frac{2a}{1+a}}$ is negative definite for any $\rho < \frac{\hat{c}_1}{3\hat{c}_2}$. By Theorem 1 in [1], the system (8) is finite-time stable. In particular, there exists a finite settling-time function

$$\bar{T}_1(y(0)) \le \frac{3\hat{c}_2(a+1)V(y(0))^{1-\frac{2a}{1+a}}}{\hat{c}_1(1-a)},$$

such that $\lim_{t\to \overline{T}_1(y(0))} y(t) = 0$ and $y(t) = 0, t > \overline{T}_1(y(0)),$ $\forall y(0) \in \mathbb{R}^{Nn}$. Under Assumption 2.1, we have $\overline{\eta} = (H^{-1} \otimes I_n)y$ and hence (14) for $T_1(\eta(0)) = \overline{T}_1((H \otimes I_n)\overline{\eta}(0)).$

IV. ROBUST CONTROLLER DESIGN

Based on the fixed-time observer (5), we further propose a distributed robust control law to solve the leader-following fixed-time synchronization problem for the multiple Euler-Lagrange systems. It is assumed the model (1) contains uncertainties and the terms $M_i(q_i)$, $C_i(q_i, \dot{q}_i)$, and $G_i(q_i)$ are not completely known, but they satisfy the following bounded conditions

$$k_{\underline{m}}I_{\underline{m}} \leq M_i(q_i) \leq k_{\overline{m}}I_{\underline{m}}, \|C_i(q_i, \dot{q}_i)\| \leq k_c \|\dot{q}_i\|, \ \|G_i(q_i)\| \leq k_g, \ \forall q_i, \dot{q}_i \in \mathbb{R}^m,$$
(15)

for some positive constants $k_{\underline{m}}$, $k_{\overline{m}}$, k_c and k_g . Throughout the section, we consider every individual agent $i = 1, \dots, N$.

First, the equations in (1) can be rewritten as, with $v_i = \dot{q}_i$,

$$\dot{q}_i = v_i, \ \dot{v}_i = M_i^{-1}(q_i)(\tau_i - C_i(q_i, v_i)v_i - G_i(q_i)),$$

which is a second-order system in the presence parametric uncertainty, i.e., the terms $M_i(q_i)$, $C_i(q_i, v_i)$ and $G_i(q_i)$ are unknown for controller design. Therefore, the conventional fixed-time control laws for second-order systems cannot be directly applied. To introduce a new method, we perform the following transformation:

$$\bar{q}_i = q_i - E\eta_i, \bar{v}_i = v_i - E\dot{\eta}_i, \tau_i = \hat{M}u_i, \ \hat{M} = \frac{2I_m}{k_m^{-1} + k_m^{-1}}.$$

Also, let $u_i = u_{1i} + u_{2i}$ with u_{1i} and u_{2i} to be designed. As a result, the above equations become

$$\begin{split} \dot{\bar{q}}_i &= \bar{v}_i, \ \dot{\bar{v}}_i = M_i^{-1}(q_i)\tau_i + F_i(q_i, v_i) - E\ddot{\eta}_i \\ &= u_{1i} + u_{2i} + (M_i^{-1}(q_i)\hat{M} - I_m)(u_{1i} + u_{2i}) \\ &+ F_i(q_i, v_i) - E\ddot{\eta}_i \end{split}$$

where $F_i(q_i, v_i) = -M_i^{-1}(q_i)(C_i(q_i, v_i)v_i + G_i(q_i))$. Moreover, it can be put in a compact form

$$\dot{\bar{q}}_i = \bar{v}_i, \quad \dot{\bar{v}}_i = u_{2i} - E\ddot{\eta}_i + Z_i \tag{16}$$

with

$$Z_i = u_{1i} + (M_i^{-1}(q_i)\hat{M} - I_m)(u_{1i} + u_{2i}) + F_i(q_i, v_i).$$
(17)

Inspired by [13], we construct a lemma that motivates the design of u_{1i} .

Lemma 4.1: Consider the quantity Z_i defined in (17) with the control law

$$u_{1i} = \begin{cases} -\frac{\kappa}{1-\epsilon} \frac{\zeta_i}{\|\zeta_i\|} (\epsilon \|u_{2i}\| + f_i(v_i)), & \|\zeta_i\| \neq 0\\ 0, & \|\zeta_i\| = 0 \end{cases}$$

$$\kappa \ge 1, \ \epsilon = \frac{k_m^{-1} - k_m^{-1}}{k_m^{-1} + k_m^{-1}}, \ f_i(v_i) = k_m^{-1} (k_c \|v_i\|^2 + k_g), \qquad (18)$$

for an arbitrary $\zeta_i \in \mathbb{R}^m$. Then, $\zeta_i^T Z_i \leq 0$ holds for any $u_{2i} \in \mathbb{R}^m$. *Proof:* From the properties of Euler-Lagrange system, we have the following facts:

$$\|M_i^{-1}(q_i)\hat{M} - I_m\| = \|\frac{2M_i^{-1}}{k_m^{-1} + k_m^{-1}} - I_m\| \le \epsilon$$
$$\|F_i(q_i, v_i)\| \le k_m^{-1}(k_c \|v_i\|^2 + k_g) = f_i(v_i),$$

which will be used in the calculation below.

When $\|\zeta_i\| = 0$, $\zeta_i^T Z_i(t) \le 0$ holds trivially. Otherwise, one has the following direct calculation

$$\begin{aligned} \zeta_i^T Z_i &\leq \zeta_i^T u_{1i} + \|\zeta_i\|(\|M_i^{-1}(q_i)\hat{M} - I_m\|(\|u_{1i}\| + \|u_{2i}\|) \\ &+ \|F_i(q_i, v_i)\|) \\ &\leq \zeta_i^T u_{1i} + \|\zeta_i\|(\epsilon(\|u_{1i}\| + \|u_{2i}\|) + f_i(v_i)) \\ &\leq (-\frac{\kappa}{1-\epsilon} + 1 + \frac{\epsilon\kappa}{1-\epsilon})\|\zeta_i\|(\epsilon\|u_{2i}\| + f_i(v_i)) \\ &= -(\kappa - 1)\|\zeta_i\|(\epsilon\|u_{2i}\| + f_i(v_i)) \leq 0, \end{aligned}$$

which completes the proof.

Remark 4.1: As the system dynamics considered in this paper contain uncertainties, a robust control approach is used in the design of u_{1i} . In particular, to guarantee $\zeta_i^T Z_i(t) \leq 0$, which will be used later for proof of convergence, u_{1i} is designed based on the boundaries of the uncertainties characterized by (15) via high gain domination. It is worth noting that the control gains in u_{1i} become higher if k_c and k_g are larger and/or ϵ is closer to 1 (corresponding to a bigger difference between $k_{\underline{m}}$ and $k_{\overline{m}}$), i.e., the size of uncertainties is larger. It is a general principle in robust control that control gains depend on the size of uncertainties. In practice, when system parameters cannot be precisely measured, a smaller range of uncertainties would be beneficial for controller design.

With Lemma 4.1 ready for u_{1i} , the remaining task is to select a specific ζ_i and design u_{2i} such that the second-order system (16) is fixed-time stable, which is more complicated than finite-time control; see some existing methods in [5], [6], [21]. For solving such problem, we first introduce an explicit procedure of designing a set of parameters to be used for the controller design. It is worth noting that these parameters are independent of system dynamics. Let $\frac{1}{2} < \alpha < 1$ and $\beta > 1$ be two specified rational numbers of ratio of two odd integers. Define four constants

$$p_1 = 0, \ p_2 = \beta - \alpha, \ p_3 = \frac{\beta}{\alpha} - \beta + \alpha - 1, \ p_4 = \frac{\beta}{\alpha} - 1$$

and four functions, for $p \ge 0$ and $\lambda > 0$,

$$l_{1}(p) = \frac{2^{1-\alpha}p}{p+1+\alpha} + \frac{p+\alpha}{p+\alpha+1}, \ l_{2}(p) = \frac{p+\beta}{p+\beta+1},$$

$$l_{3}(p,\lambda) = \frac{2^{1-\alpha}(1+\alpha)}{p+1+\alpha}\lambda^{\frac{p+\alpha+1}{1+\alpha}} + \frac{(\lambda\gamma_{1})^{p+\alpha+1}}{p+\alpha+1},$$

$$l_{4}(p,\lambda) = \frac{(\lambda\gamma_{2})^{p+\beta+1}}{p+\beta+1}.$$

For the convenience of presentation, we also define

$$\ell_1(p) = (2 - \alpha)2^{1-\alpha}l_1(p), \ \ell_2(p) = (2 - \alpha)2^{1-\alpha}l_2(p), \ell_3(p,\lambda) = (2 - \alpha)2^{1-\alpha}l_3(p,\lambda), \ \ell_4(p,\lambda) = (2 - \alpha)2^{1-\alpha}l_4(p,\lambda).$$

Next, pick $L_1 = \max\{\ell_2(p_2), \ell_1(p_3)\}$ and two positive parameters

$$\gamma_{1} > \max\left\{\frac{2^{1-\alpha}}{1+\alpha} + \ell_{1}(p_{1}) + 2L_{1}, \frac{2^{1-\alpha}\frac{\beta}{\alpha}}{\frac{\beta}{\alpha}+\alpha} + \ell_{2}(p_{3}) + \ell_{1}(p_{4})\right\}$$
$$\gamma_{2} > \max\left\{\ell_{2}(p_{4}) + 2L_{1}, \ell_{2}(p_{1}) + \ell_{1}(p_{2})\right\}.$$
(19)

Now, it is ready to select

$$\lambda_1 = \gamma_1^{\frac{1}{\alpha}}, \ \lambda_2 = \gamma_1^{\frac{1}{\alpha}-1} \frac{\gamma_2 \beta}{\alpha}, \ \lambda_3 = \gamma_1 \gamma_2^{\frac{1}{\alpha}-1}, \ \lambda_4 = \gamma_2^{\frac{1}{\alpha}} \frac{\beta}{\alpha}.$$

Then, pick

$$L_{2} = \max\left\{\frac{2^{1-\alpha}\alpha}{\frac{\beta}{\alpha}+\alpha} + \ell_{4}(p_{3},\lambda_{3}) + \ell_{3}(p_{4},\lambda_{4}), \\ \ell_{4}(p_{1},\lambda_{1}) + \ell_{3}(p_{2},\lambda_{2}), \ell_{4}(p_{2},\lambda_{2}), \ell_{3}(p_{3},\lambda_{3})\right\}$$

Finally, we select the following two parameters

$$k_1 > \frac{2^{1-\alpha}\alpha}{\alpha+1} + \ell_3(p_1,\lambda_1) + 4L_2, \ k_2 > \ell_4(p_4,\lambda_4) + 4L_2.$$
 (20)

With these parameters obtained, it is ready to have the following lemma.

Lemma 4.2: Consider the system (16) where u_{1i} is given in (18) with

$$\zeta_i = \varepsilon_i^{2-\alpha} \tag{21}$$

and

$$u_{2i} = -k_1 \varepsilon_i^{2\alpha - 1} - k_2 \varepsilon_i^{\frac{\beta}{\alpha} + \beta + \alpha - 2} + ESS\eta_i,$$

$$\varepsilon_i = \bar{v}_i^{\frac{1}{\alpha}} + (\gamma_1 \bar{q}_i^{\alpha} + \gamma_2 \bar{q}_i^{\beta})^{\frac{1}{\alpha}}.$$
(22)

Suppose the observer governing η_i satisfies Lemma 3.1. If the control parameters $\gamma_1, \gamma_2, k_1, k_2$ satisfy (19) and (20), then the equilibrium of (16) at the origin is fixed-time stable. In particular, there exists a constant settling-time $T_2^* \ge 0$ such that

$$\lim_{t \to T_1^* + T_2^*} [\bar{q}_i(t), \bar{v}_i(t)] = 0,$$

$$[\bar{q}_i(t), \bar{v}_i(t)] = 0, \ t \ge T_1^* + T_2^*, \ \forall \bar{q}_i(T_1^*), \bar{v}_i(T_1^*) \in \mathbb{R}^m.$$
(23)

Proof: For the convenience of proof, we define the following variables

$$\bar{\varepsilon}_i = \varepsilon_i^{2\alpha - 1}, \quad \hat{\varepsilon}_i = \varepsilon_i^{\frac{\beta}{\alpha} + \beta + \alpha - 2} \\ \hat{v}_i^* = -\gamma_1 \bar{q}_i^{\alpha} - \gamma_2 \bar{q}_i^{\beta}, \quad \varepsilon_i = \bar{v}_i^{\frac{1}{\alpha}} - \hat{v}_i^{*\frac{1}{\alpha}}$$

Let

$$W_{1i}(\bar{q}_i) = \frac{1}{2} \|\bar{q}_i\|^2 + \frac{1}{\beta/\alpha + 1} \|\bar{q}_i^{\frac{\beta}{\alpha} + 1}\|_1.$$

Along the trajectory of \bar{q}_i —th subsystem in (16), the derivative of $W_{1i}(\bar{q}_i)$ satisfies

$$\begin{split} \dot{W}_{1i}(\bar{q}_i) = & (\bar{q}_i + \bar{q}_i^{\frac{\beta}{\alpha}})^T \bar{v}_i = (\bar{q}_i + \bar{q}_i^{\frac{\beta}{\alpha}})^T (\bar{v}_i - \hat{v}_i^* + \hat{v}_i^*) \\ = & (\bar{q}_i + \bar{q}_i^{\frac{\beta}{\alpha}})^T (\bar{v}_i - \hat{v}_i^*) - (\bar{q}_i + \bar{q}_i^{\frac{\beta}{\alpha}})^T (\gamma_1 \bar{q}_i^{\alpha} + \gamma_2 \bar{q}_i^{\beta}). \end{split}$$

By Lemma 2.2, for $0 < \alpha < 1$, one has

$$\begin{aligned} |\bar{v}_i - \hat{v}_i^*| &= |(\bar{v}_i^{\frac{1}{\alpha}})^{\alpha} - (\hat{v}_i^{*\frac{1}{\alpha}})^{\alpha}| \\ \leq 2^{1-\alpha} |\bar{v}_i^{\frac{1}{\alpha}} - \hat{v}_i^{*\frac{1}{\alpha}}|^{\alpha} &= 2^{1-\alpha} |\varepsilon_i|^{\alpha}. \end{aligned}$$
(24)

And, by Lemma 2.3,

$$(\bar{q}_{i} + \bar{q}_{i}^{\frac{\alpha}{\alpha}})^{T}(\bar{v}_{i} - \hat{v}_{i}^{*}) \leq 2^{1-\alpha} |\bar{q}_{i}|^{T} |\varepsilon_{i}|^{\alpha} + 2^{1-\alpha} |\bar{q}_{i}^{\frac{\beta}{\alpha}}|^{T} |\varepsilon_{i}|^{\alpha}$$

$$\leq 2^{1-\alpha} (\frac{1}{1+\alpha} \|\bar{q}_{i}^{\alpha+1}\|_{1} + \frac{\alpha}{1+\alpha} \|\varepsilon_{i}^{\alpha+1}\|_{1})$$

$$+ 2^{1-\alpha} (\frac{\frac{\beta}{\alpha}}{\frac{\beta}{\alpha}+\alpha} \|\bar{q}_{i}^{\frac{\beta}{\alpha}+\alpha}\|_{1} + \frac{\alpha}{\frac{\beta}{\alpha}+\alpha} \|\varepsilon_{i}^{\frac{\beta}{\alpha}+\alpha}\|_{1}).$$
(25)

Using (24) and (25), one has

$$\dot{W}_{1i}(\bar{q}_i) \leq -\left(\gamma_1 - \frac{2^{1-\alpha}}{1+\alpha}\right) \|\bar{q}_i^{\alpha+1}\|_1 - \gamma_2 \|\bar{q}_i^{\beta+1}\|_1 -\left(\gamma_1 - \frac{2^{1-\alpha}\frac{\beta}{\alpha}}{\frac{\beta}{\alpha}+\alpha}\right) \|\bar{q}_i^{\frac{\beta}{\alpha}+\alpha}\|_1 - \gamma_2 \|\bar{q}_i^{\frac{\beta}{\alpha}+\beta}\|_1 + \frac{2^{1-\alpha}\alpha}{\alpha+1} \|\varepsilon_i^{\alpha+1}\|_1 + \frac{2^{1-\alpha}\alpha}{\frac{\beta}{\alpha}+\alpha} \|\varepsilon_i^{\frac{\beta}{\alpha}+\alpha}\|_1.$$
(26)

Next, we define a vector function

$$d(\bar{q}_i, \bar{v}_i) = \int_{\hat{v}_i^*}^{\bar{v}_i} (s^{\frac{1}{\alpha}} - \hat{v}_i^{*\frac{1}{\alpha}})^{2-\alpha} ds.$$

and hence

$$W_{2i}(\bar{q}_i, \bar{v}_i) = \|d(\bar{q}_i, \bar{v}_i)\|_1.$$

Before the analysis on its derivative, we give the following calculation in order:

$$\int_{\hat{v}_{i}^{*}}^{\bar{v}_{i}} (s^{\frac{1}{\alpha}} - \hat{v}_{i}^{*\frac{1}{\alpha}})^{1-\alpha} ds \leq \operatorname{diag}(|\bar{v}_{i} - \hat{v}_{i}^{*}|)|\varepsilon_{i}|^{1-\alpha}, |\bar{v}_{i}|^{T} \operatorname{diag}(|\bar{v}_{i} - \hat{v}_{i}^{*}|)|\varepsilon_{i}|^{1-\alpha} \leq 2^{1-\alpha}|\bar{v}_{i}|^{T}|\varepsilon_{i}|.$$
(27)

Then, the derivative of $W_{2i}(\bar{q}_i, \bar{v}_i)$ along the trajectory of (16) satisfies, using (27),

$$\dot{W}_{2i}(\bar{q}_i, \bar{v}_i) = (2 - \alpha) \dot{\bar{q}}_i^T \frac{-\partial \bar{v}_i^{*\frac{1}{\alpha}}}{\partial \bar{q}_i} \int_{\hat{v}_i^*}^{\bar{v}_i} (s^{\frac{1}{\alpha}} - \hat{v}_i^{*\frac{1}{\alpha}})^{1-\alpha} ds + (\varepsilon_i^{2-\alpha})^T \dot{\bar{v}}_i$$
$$\leq A_1 + A_2 \tag{28}$$

for

$$A_1 = (2 - \alpha) 2^{1-\alpha} |\bar{v}_i|^T \left| \frac{\partial(\hat{v}_i^{*\frac{1}{\alpha}})}{\partial \bar{q}_i} \right| |\varepsilon_i$$
$$A_2 = (\varepsilon_i^{2-\alpha})^T (u_{2i} - E\ddot{\eta}_i + Z_i).$$

By Lemma 2.1, we can obtain

$$\begin{aligned} & \left| \frac{\partial (\hat{v}_i^{*\frac{1}{\alpha}})}{\partial \bar{q}_i} \right| \\ = & \left| \operatorname{diag}((\gamma_1 \bar{q}_i^{\alpha} + \gamma_2 \bar{q}_i^{\beta})^{\frac{1}{\alpha} - 1}) \operatorname{diag}(\gamma_1 \bar{q}_i^{\alpha - 1} + \frac{\gamma_2 \beta}{\alpha} \bar{q}_i^{\beta - 1}) \right| \\ \leq & \gamma_1^{\frac{1}{\alpha}} I_m + \gamma_1^{\frac{1}{\alpha} - 1} \frac{\gamma_2 \beta}{\alpha} \operatorname{diag}(|\bar{q}_i^{\beta - \alpha}|) \\ & + \gamma_1 \gamma_2^{\frac{1}{\alpha} - 1} \operatorname{diag}(|\bar{q}_i^{\frac{\beta}{\alpha} - \beta + \alpha - 1}|) + \gamma_2^{\frac{1}{\alpha}} \frac{\beta}{\alpha} \operatorname{diag}(|\bar{q}_i^{\frac{\beta}{\alpha} - 1}|) \end{aligned}$$

that implies $A_1 \leq (2-\alpha)2^{1-\alpha} \sum_{j=1}^4 \lambda_j |\bar{v}_i|^T \operatorname{diag}(|\bar{q}_i^{p_j}|)|\varepsilon_i|$. To simplify the presentation, we introduce the following operator

$$\langle x, y \rangle_q := x \|\bar{q}_i^y\|_1, \ \langle x, y \rangle_\varepsilon := x \|\varepsilon_i^y\|_1.$$

Then, using Lemma 2.3 and a similar argument as (24) gives

$$\begin{split} &\lambda |\bar{v}_i|^{I} \operatorname{diag}(|\bar{q}_i^{p}|)|\varepsilon_i| \\ &\leq \lambda |\bar{v}_i - \hat{v}_i^{*}|^{T} \operatorname{diag}(|\bar{q}_i^{p}|)|\varepsilon_i| + \lambda |\hat{v}_i^{*}|^{T} \operatorname{diag}(|\bar{q}_i^{p}|)|\varepsilon_i| \\ &\leq \lambda 2^{1-\alpha} |\varepsilon_i^{\alpha+1}|^{T} |\bar{q}_i|^{p} + \lambda |\gamma_1 \bar{q}_i^{\alpha} + \gamma_2 \bar{q}_i^{\beta} |\operatorname{diag}(|\bar{q}_i^{p}|)|\varepsilon_i| \\ &\leq \lambda 2^{1-\alpha} |\varepsilon_i^{\alpha+1}|^{T} |\bar{q}_i|^{p} + \lambda \gamma_1 |\varepsilon_i|^{T} |\bar{q}_i^{p+\alpha}| + \lambda \gamma_2 |\varepsilon_i|^{T} |\bar{q}_i^{p+\beta}| \\ &\leq \langle l_1(p), p + \alpha + 1 \rangle_q + \langle l_2(p), p + \beta + 1 \rangle_q \\ &+ \langle l_3(p, \lambda), p + \alpha + 1 \rangle_{\varepsilon} + \langle l_4(p, \lambda), p + \beta + 1 \rangle_{\varepsilon} \end{split}$$

and hence

$$A_{1} \leq \sum_{j=1}^{4} \langle \ell_{1}(p_{j}), p_{j} + \alpha + 1 \rangle_{q} + \langle \ell_{2}(p_{j}), p_{j} + \beta + 1 \rangle_{q} + \langle \ell_{3}(p_{j}, \lambda_{j}), p_{j} + \alpha + 1 \rangle_{\varepsilon} + \langle \ell_{4}(p_{j}, \lambda_{j}), p_{j} + \beta + 1 \rangle_{\varepsilon}.$$
(29)

By Lemma 3.1, $y_i(t) = 0$ and hence $SS\eta_i - \ddot{\eta}_i = 0$ for $t \ge T_1^*$. By Lemma 4.1, one has $\zeta_i^T Z_i \le 0$. Since $|\varepsilon_{ji}^{\alpha+1}| = \varepsilon_{ji}^{\alpha+1}$ and $|\varepsilon_{ji}^{\frac{\alpha}{\alpha}+\beta}| = \varepsilon_{ji}^{\frac{\beta}{\alpha}+\beta}$, $j = 1, \dots, m$, for α and β being two rational numbers of ratio of two odd integers and ε_{ji} being the *j*-th entry of ε_i ,

$$(\varepsilon_i^{2-\alpha})^T \varepsilon_i^{2\alpha-1} = \sum_{j=1}^m \varepsilon_{ji}^{\alpha+1} = \sum_{j=1}^m |\varepsilon_{ji}^{\alpha+1}| = \|\varepsilon_i^{\alpha+1}\|_1,$$
$$(\varepsilon_i^{2-\alpha})^T \varepsilon_i^{\frac{\beta}{\alpha}+\beta+\alpha-2} = \|\varepsilon_i^{\frac{\beta}{\alpha}+\beta}\|_1.$$

Thus,

$$A_{2} = (\varepsilon_{i}^{2-\alpha})^{T} (-k_{1}\varepsilon_{i}^{2\alpha-1} - k_{2}\varepsilon_{i}^{\frac{\beta}{\alpha}+\beta+\alpha-2}) + (\varepsilon_{i}^{2-\alpha})^{T} (ESS\eta_{i} - E\ddot{\eta}_{i}) + \zeta_{i}^{T}Z_{i} \leq -k_{1} \|\varepsilon_{i}^{\alpha+1}\|_{1} - k_{2} \|\varepsilon_{i}^{\frac{\beta}{\alpha}+\beta}\|_{1}.$$
(30)

Let

$$W_i(\bar{q}_i, \bar{v}_i) = W_{1i}(\bar{q}_i) + W_{2i}(\bar{q}_i, \bar{v}_i).$$

Under the conditions for γ_1 , γ_2 , k_1 , and k_2 , there exist $\hat{\gamma} > 0$ and $\hat{k} > 0$ satisfying

$$\gamma_1 - \frac{2^{1-\alpha}}{1+\alpha} - \ell_1(p_1) \ge \hat{\gamma} + 2L_1, \ \gamma_2 - \ell_2(p_4) \ge \hat{\gamma} + 2L_1$$

and

$$k_1 - \frac{2^{1-\alpha}\alpha}{\alpha+1} - \ell_3(p_1,\lambda_1) \ge \hat{k} + 4L_2, \ k_2 - \ell_4(p_4,\lambda_4) \ge \hat{k} + 4L_2.$$

Then, combining (26), (28), (29), and (30) gives, for $t \ge T_1^*$,

$$\dot{W}_i(\bar{q}_i, \bar{v}_i) \le -B_1 - B_2$$

where

$$\begin{split} B_1 = & \langle \hat{\gamma} + 2L_1, \alpha + 1 \rangle_q + \langle \hat{\gamma} + 2L_1, \frac{\beta}{\alpha} + \beta \rangle_q \\ & - \langle L_1, 2\beta - \alpha + 1 \rangle_q - \langle L_1, \frac{\beta}{\alpha} - \beta + 2\alpha \rangle_q \\ B_2 = & \langle \hat{k} + 4L_2, \alpha + 1 \rangle_{\varepsilon} + \langle \hat{k} + 4L_2, \frac{\beta}{\alpha} + \beta \rangle_{\varepsilon} \\ & - \langle L_2, \frac{\beta}{\alpha} + \alpha \rangle_{\varepsilon} - \langle L_2, \beta + 1 \rangle_{\varepsilon} \\ & - \langle L_2, 2\beta - \alpha + 1 \rangle_{\varepsilon} - \langle L_2, \frac{\beta}{\alpha} - \beta + 2\alpha \rangle_{\varepsilon}. \end{split}$$

It is easy to verify the following inequalities

$$\begin{aligned} \frac{\beta}{\alpha} + \beta &> 2\beta - \alpha + 1 > \alpha + 1 \\ \frac{\beta}{\alpha} + \beta &> \frac{\beta}{\alpha} - \beta + 2\alpha > \alpha + 1 \\ \frac{\beta}{\alpha} + \beta &> \frac{\beta}{\alpha} + \alpha > \alpha + 1 \\ \frac{\beta}{\alpha} + \beta &> \beta + 1 > \alpha + 1. \end{aligned}$$

Therefore,

$$\langle 2L_1, \alpha + 1 \rangle_q + \langle 2L_1, \frac{\beta}{\alpha} + \beta \rangle_q \geq \langle L_1, 2\beta - \alpha + 1 \rangle_q + \langle L_1, \frac{\beta}{\alpha} - \beta + 2\alpha \rangle_q$$

that implies $B_1 \geq \langle \hat{\gamma}, \alpha + 1 \rangle_q + \langle \hat{\gamma}, \frac{\beta}{\alpha} + \beta \rangle_q$. Similarly, one has $B_2 \geq \langle \hat{k}, \alpha + 1 \rangle_{\varepsilon} + \langle \hat{k}, \frac{\beta}{\alpha} + \beta \rangle_{\varepsilon}$. The above two inequalities conclude

$$\dot{W}_{i}(\bar{q}_{i},\bar{v}_{i}) \leq -\langle\hat{\gamma},\alpha+1\rangle_{q} - \langle\hat{\gamma},\frac{\beta}{\alpha}+\beta\rangle_{q} -\langle\hat{k},\alpha+1\rangle_{\varepsilon} - \langle\hat{k},\frac{\beta}{\alpha}+\beta\rangle_{\varepsilon}.$$
(31)

Next, since

$$\|d(\bar{q}_i,\bar{v}_i)\|_1 \leq \operatorname{diag}(|\bar{v}_i-\hat{v}_i^*|)|\varepsilon_i|^{2-\alpha} \leq 2^{1-\alpha}\|\varepsilon_i\|^2,$$

one has

$$W_i(\bar{q}_i, \bar{v}_i) \le \frac{1}{2} \|\bar{q}_i\|^2 + \frac{1}{\beta/\alpha + 1} \|\bar{q}_i^{\frac{\beta}{\alpha} + 1}\|_1 + 2^{1-\alpha} \|\varepsilon_i\|^2.$$

Direct calculation, using Lemma 2.2, gives

$$W_i^{\frac{\alpha+1}{2}} \leq \langle \nu_1, \alpha+1 \rangle_q + \langle \nu_1, (\frac{\beta}{\alpha}+1)\frac{\alpha+1}{2} \rangle_q + \langle \nu_1, \alpha+1 \rangle_{\varepsilon}$$

and

$$W_i^{\frac{\frac{\beta}{\alpha}+\beta}{\frac{\beta}{\alpha}+1}} \leq \langle \nu_2, \frac{2(\frac{\beta}{\alpha}+\beta)}{\frac{\beta}{\alpha}+1} \rangle_q + \langle \nu_2, \frac{\beta}{\alpha}+\beta \rangle_q + \langle \nu_2, \frac{2(\frac{\beta}{\alpha}+\beta)}{\frac{\beta}{\alpha}+1} \rangle_{\varepsilon}$$

for some constants $\nu_1, \nu_2 > 0$. Again, It is easy to verify the following inequalities

$$\frac{\beta}{\alpha} + \beta > (\frac{\beta}{\alpha} + 1)\frac{\alpha + 1}{2} > \alpha + 1$$
$$\frac{\beta}{\alpha} + \beta > \frac{2(\frac{\beta}{\alpha} + \beta)}{\frac{\beta}{\alpha} + 1} > \alpha + 1.$$

Therefore,

$$W_{i}^{\frac{\alpha+1}{2}} \leq \langle 2\nu_{1}, \alpha+1 \rangle_{q} + \langle \nu_{1}, \frac{\beta}{\alpha} + \beta \rangle_{q} + \langle \nu_{1}, \alpha+1 \rangle_{\varepsilon}$$
$$W_{i}^{\frac{\beta}{\alpha} + \beta} \leq \langle \nu_{2}, \alpha+1 \rangle_{q} + \langle 2\nu_{2}, \frac{\beta}{\alpha} + \beta \rangle_{q}$$
$$+ \langle \nu_{2}, \alpha+1 \rangle_{\varepsilon} + \langle \nu_{2}, \frac{\beta}{\alpha} + \beta \rangle_{\varepsilon}.$$
(32)

Comparing (31) with (32), one can conclude

$$\dot{W}_i \leq -\rho_1 W_i^{\frac{\alpha+1}{2}} - \rho_2 W_i^{\frac{\frac{\beta}{\alpha}+\beta}{\frac{\beta}{\alpha}+1}}$$

for $\rho_1 = \min\left\{\frac{\hat{\gamma}}{2\nu_1}, \frac{\hat{\kappa}}{\nu_1}\right\}/2$, $\rho_2 = \min\left\{\frac{\hat{\gamma}}{2\nu_2}, \frac{\hat{\kappa}}{\nu_2}\right\}/2$. By Lemma 2.4, the equilibrium of (16) is fixed-time stable. In particular, there exists

$$T_2^* \le \frac{2}{\rho_1(1-\alpha)} + \frac{\beta + \alpha}{\rho_2 \alpha(\beta - 1)}$$

such that (23) holds.

Finally, based on Lemma 3.1 and Lemma 4.2, we can obtain the following theorem for the solvability of the fixed-time synchronization problem with $T^* = T_1^* + T_2^*$.

Theorem 4.1: The fixed-time synchronization problem for the multi-agent system composed of (1) and (2) under Assumption 2.1 is solvable by the observer (5) and the controller $\tau_i = \hat{M}(u_{1i} + u_{2i})$ of the form (18) and (22) with all the parameters given in Lemma 3.1 and Lemma 4.2.

Remark 4.2: Suppose the sub-controller u_{1i} follows (18) with (21) but the sub-controller (22) for u_{2i} reduces to the following finite-time controller, by setting $k_2 = 0$ and $\gamma_2 = 0$,

$$u_{2i} = -k_1 \varepsilon_i^{2\alpha - 1} + ESS\eta_i, \quad \varepsilon_i = \bar{v}_i^{\frac{1}{\alpha}} + \gamma_1^{\frac{1}{\alpha}} \bar{q}_i, \tag{33}$$

where η_i is governed by the finite-time observer (13). If $\frac{1}{2} < \alpha < 1$ is a ratio of two odd integers and γ_1, k_1 satisfy

$$\begin{split} \gamma_1 &> \frac{2^{1-\alpha}}{1+\alpha} + \frac{\alpha(2-\alpha)2^{1-\alpha}}{1+\alpha} \\ k_1 &> \gamma_1^{1+1/\alpha} \left(\frac{2^{1-\alpha}\alpha}{1+\alpha} + \frac{(2-\alpha)2^{1-\alpha}}{\gamma_1} (2^{1-\alpha} + \frac{\gamma_1}{1+\alpha}) \right), \end{split}$$

then the equilibrium of the closed-loop system composed of (16) at the origin is finite-time stable. In particular, there exists a finite settling-time function $T_{2i}(\bar{q}_i(T_1(\eta(0))), \bar{v}_i(T_1(\eta(0)))) \ge 0$ such that

$$\lim_{t \to T_1 + T_{2i}} [\bar{q}_i(t), \bar{v}_i(t)] = 0,$$

$$[\bar{q}_i(t), \bar{v}_i(t)] = 0, \ t \ge T_1 + T_{2i}, \ \forall \bar{q}_i(T_1), \bar{v}_i(T_1) \in \mathbb{R}^m.$$
(34)

As a result, the finite-time synchronization problem for the multiagent system composed of (1) and (2) under Assumption 2.1 is solvable by the observer (13) and the controller $\tau_i = \hat{M}(u_{1i} + u_{2i})$ of the form (18) and (33). The proof can similarly follow that of Lemma 4.2 and is thus omitted.

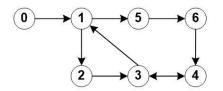


Fig. 1. Illustration of the communication network topology.

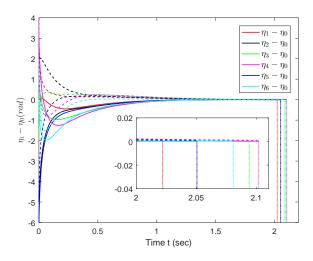


Fig. 2. Profile of the estimation errors $\eta_i - \eta_0$, $i = 1, \dots, 6$, under the fixed-time observer.

V. AN EXAMPLE

Consider a group of six robotic manipulators given by (1) where $q_i = [q_{1i}, q_{2i}]^T \in \mathbb{R}^2$ and

$$\begin{split} M_{i}(q_{i}) &= \begin{bmatrix} \theta_{1i} + \theta_{2i} + 2\theta_{3i}\cos(q_{2i}) & \theta_{2i} + \theta_{3i}\cos(q_{2i}) \\ \theta_{2i} + \theta_{3i}\cos(q_{2i}) & \theta_{4i} \end{bmatrix} \\ C_{i}(q_{i}, \dot{q}_{i})\dot{q}_{i} &= \begin{bmatrix} -\theta_{3i}\sin(q_{2i})\dot{q}_{1i}^{2} - 2\theta_{3i}\sin(q_{2i})\dot{q}_{1i}\dot{q}_{2i} \\ \theta_{3i}\sin(q_{2i})\dot{q}_{2i}^{2} \end{bmatrix} \\ G_{i}(q_{i}) &= \begin{bmatrix} \theta_{5i}g\cos(q_{1i}) + \theta_{6i}g\cos(q_{1i} + q_{2i}) \\ \theta_{6i}g\cos(q_{1i} + q_{2i}) \end{bmatrix} \end{split}$$

for $i = 1, \dots, 6$. In the equations, $\theta_{ji}, j = 1, \dots, 6, i = 1, \dots, 6$, represent unknown parameters. The leader system is given by (2) with $S = \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix}$ and $E = I_2$. The information flow among all the subsystems and the leader is described by the digraph in Fig. 1, which contains a spanning tree with node 0 as the root, satisfying Assumption 2.1. Let $D = 8I_6$. Then $DH + H^T D \ge 2I_N$.

By Lemma 3.1, let $c_1 = 8.4$, $c_2 = 1$, $c_3 = 1$, a = 3/5, b = 3. Now we can construct the fixed-time observer (5) whose performance is shown in Fig. 2. It is observed that the estimation errors $\eta_i - \eta_0$, $i = 1, \dots, 6$, approach zero at the time instants marked by the vertical lines. In the simulation, the error tolerance of numerical calculation is set as 10^{-3} that is used as the criterion of approaching zero.

Next, we apply the observer (5) to solve the fixed-time control problem of Euler-Lagrange systems and design the fixed-time control law $\tau_i = \hat{M}(u_{1i} + u_{2i})$ where u_{1i} is given by (18) and u_{2i} is given by (22). Although we we do not know the exact value of $M_i(q_i)$, $C_i(q_i, \dot{q}_i)$ and $G_i(q_i)$, it is assumed that the unknown parameters in the following ranges $\theta_{1i} \in [6, 8]$, $\theta_{2i} \in [0.8, 1]$, $\theta_{3i} \in [1, 1.4]$, $\theta_{5i} \in [1.5, 2]$, and $\theta_{6i} \in [1, 1.3]$. Simple calculation verifies that the properties in (15) are satisfied for $k_m^{-1} = 0.3$, $k_m^{-1} = 0.08$, $k_c = 3$, and $k_g = 50$. We select the parameters in (18) and (22) as $\kappa = 3$, $\epsilon =$

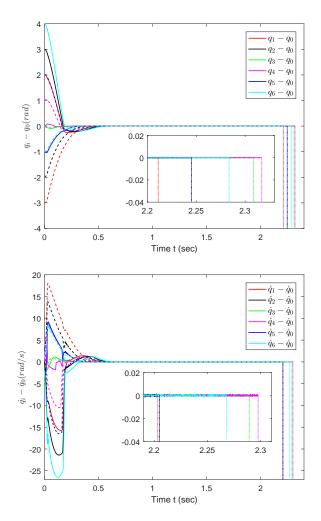


Fig. 3. Profile of the synchronization errors $q_i - q_0$ and $\dot{q}_i - \dot{q}_0$, $i = 1, \dots, 6$, under the fixed-time controller.

11/19, $\gamma_1 = 10$, $\gamma_2 = 10$, $k_1 = 20$, $k_2 = 15$, $\alpha = 7/9$, $\beta = 9/7$. For the purpose of simulation, we provide the values for uncertain parameters $\theta_{1i} = 7$, $\theta_{2i} = 0.96$, $\theta_{3i} = 1.2$, $\theta_{4i} = 5.96$, $\theta_{5i} = 2$, and $\theta_{6i} = 1.2$. The simulation is conducted with arbitrarily selected initial conditions. Fig. 3 shows q_i , \dot{q}_i respectively converge to q_0 , \dot{q}_0 in fixed time instants.

VI. CONCLUSION

This paper has proposed the fixed-time robust control design for the consensus problem of networked Euler-Lagrange systems based on a distributed observer, which is capable of estimating the desired trajectory of the leader in a fixed time under a directed graph. The heterogeneous uncertain Euler-Lagrange systems are converted into second-order systems by a partial design of the control law, and then the backstepping procedure for second-order systems are utilized to accomplished the fixed-time control design.

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